

Quasi Elliptic Bandpass Filter with Defected Ground Structure for Improved Performance

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Abstract

Filters are widely used in the telecommunication industry in recent years. They selectively sort signals and pass through a desired range of signals while suppressing or removing others. This project presents the design of a quasi-elliptic bandpass filter (QE) that operates at 2.4 GHz employing Defected Ground Structure (DGS) slot. The objective of this project is to enhance the performance of a bandpass filter in terms of better return loss and increased bandwidth by employing DGS element in the ground plane. The chosen filter topology is based on a square open loop resonator. DGS elements of rectangular geometries have been chosen for the project. Two sets of QE filters have been designed and investigated. The first set is without DGS, while the second set is with variations of rectangular DGS slot elements. The filters are named QE and QE_R filter sets. The effects of introducing DGS slots into the filter ground plane have been successfully investigated. It is observed that the QE_R filter exhibits excellent low return loss at 2.4 GHz, and increased operating bandwidth, compared to the QE filter. The optimum configuration is chosen to be the QE_2R filter. It is observed that the QE_2R filter exhibits excellent simulated low return loss of -28.07 dB at 2.4 GHz, and the bandwidth of the filter reached to 98.39 MHz which is 19.39 MHz wider than QE filter. Its stopband responses are quite similar, ie approximately -30 dB, within the observed lower and higher bands.

Keywords: quasi-elliptic bandpass filter; defected ground structure; microstrip; loop resonator; and enhanced performance

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1. Introduction

Filters are important components being widely used in the telecommunication industry. A filter selectively sorts signals and passes through a desired range of signals while blocking or suppressing others. This article presents the design of a quasi-elliptic bandpass filter, named QE filter. It operates at 2.4 GHz, and is integrated with Defected Ground Structure (DGS). The design aims for enhanced performance in terms of better return loss and increased operating bandwidth. The chosen filter topology is based on a square open loop resonator. DGS elements of rectangular geometries are integrated in the ground plane. Two sets of QE filters have been successfully designed and simulated. The first set is without DGS, while the second set is with variations of rectangular DGS slot elements. The filters are named QE and QE_R filter sets. It is observed that the QE_R filter exhibits excellent low return loss at 2.4 GHz, and increased operating bandwidth, compared to the QE filter.

2. Quasi Elliptic Filter and DGS Structures

A QE filter which consists of four square resonators shown in Figure 1 has been successfully simulated (*Mohd Akram 2009*). The main advantage of the open square resonator is its ability to provide both positive and negative coupling coefficients, but alleviate the limitation on the strength of the coupling coefficient which can be implemented.

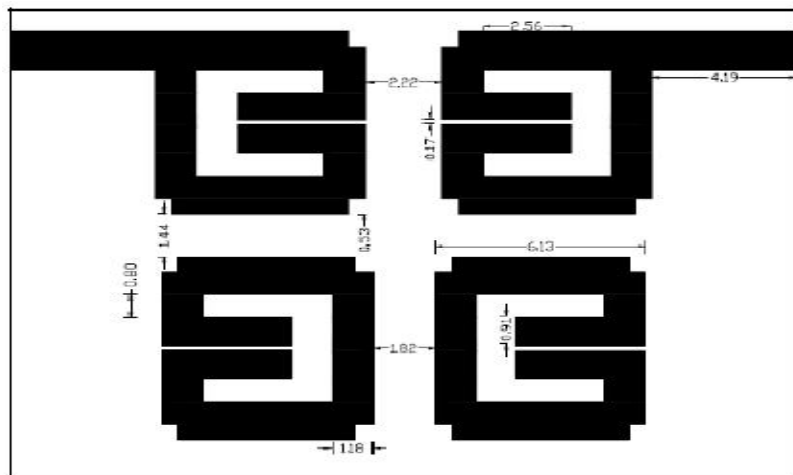


Figure 1: Geometry of QE Open Loop Resonator Filter

One recently introduced method to improve the filter characteristics is to have defected ground plane. Various slot geometries have been reported in the literature for the purpose of reducing harmonics, producing frequency pass or stop-bands, and enhancing coupling effects (A. Abdel-Rahman 2006). A DGS cell behaves as complementary to the main filter and independently in executing the filter, due to its natural resonant characteristic.

A periodic structure or slot geometry present on the ground plane can disturb the field distribution of the guided electromagnetic wave and create an electromagnetic bandgap (EBG) effect within certain frequency ranges. The effect is useful for suppressing undesired harmonics and preventing undesired signals in the components, which is the primary function of filters. The DGS structure can be realised by etching off its geometry from the conducting ground plane. Such slot in the ground plane can enhance the electric coupling, or the electric part of a mixed coupling between two adjacent resonators. The behavior can be seen in Figures 2 and 3.

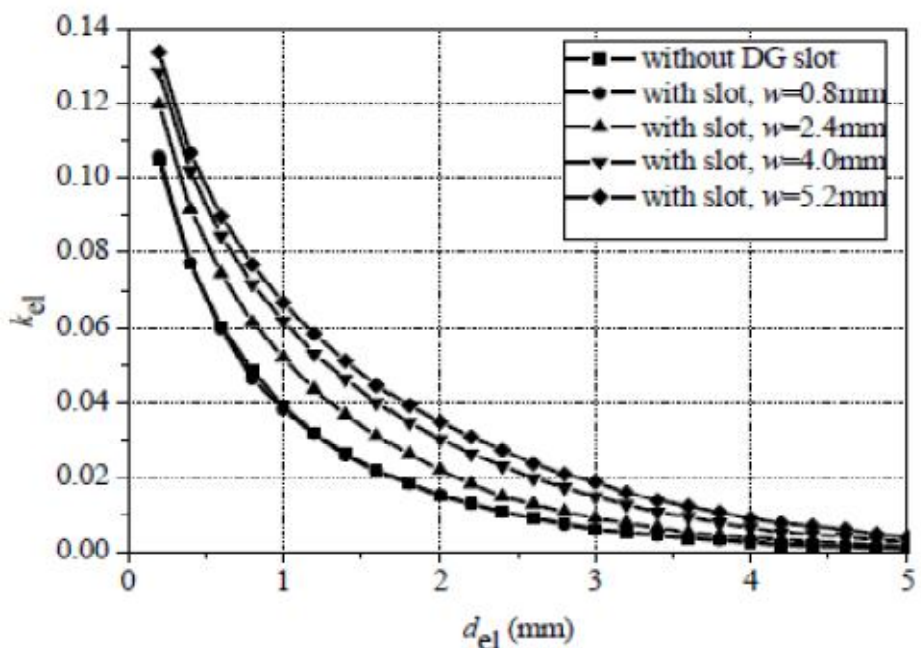


Figure 2: Electric Coupling Coefficient vs. Coupling gap (N. Militarú 2007)

Fig. 3 shows the dependence of the electric coupling coefficient on the gap between resonators, for several width values of the ground slot. The presence of slot enhances the electric coupling coefficient. The magnetic coupling coefficient is slightly larger, compared to the classical microstrip structure.

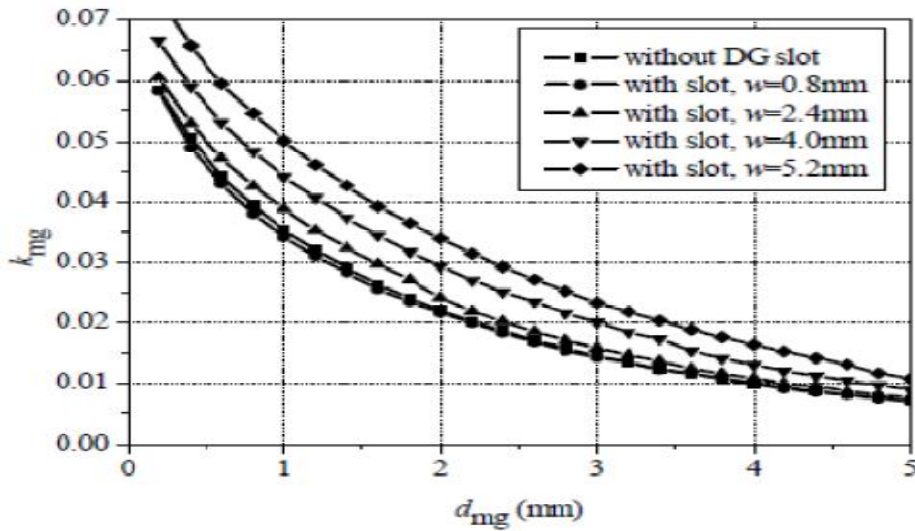


Figure 3: Magnetic Coupling Coefficient vs. Coupling Gap (N. Militaru 2007)

An etched defect in ground plane disturbs the shield current distribution that is present. The disturbance can change the characteristics of a transmission line such as line capacitance and inductance.

3. Design of Quasi-Elliptic Filter

Table 1 shows the design specification of the QE filter. Table 2 shows the properties of the low loss Rogers microwave board, RT/Duroid 6010. During the design, modifications are necessary in order to fine-tune the filter for optimised performance.

Table 1: Quasi Elliptic Filter Specification

Center frequency	2.4Ghz
Fractional bandwidth	>4%
Passband return loss	>-25dB
Passband insertion loss	-1dB
VSWR	<2
Stop band attenuation	< -20dB
Response type	Elliptic
I/O impedance	50 Ω

Table 2: RT/Duroid 6010 Material Properties

Dielectric constant	10.2
Loss tangent	0.0023
Substrate thickness	1.27 mm
Conductor thickness	35 μ m
Substrate height	0.25 mm
Dimension of the board	18 mm X 18 mm

In order to design the microstrip quasi elliptic filter using resonators, cross coupled topology was chosen. One type is using square open loop resonator. Figure 4 shows the geometry of a square resonator.



Figure 4: Square Resonator

There are several DGS shapes that have been reported in the literature such as alphabetical, hexagonal, circular and dump-bell shapes. For the design, rectangular geometry has been chosen (*M. Azfar A. Aziz 2011*). Both the OE filters, with and without DGS element are simulated and analysed. Further investigations were carried out with varying DGS sizes. The performances are then compared.

Figure 5 shows the proposed geometry of DGS element.

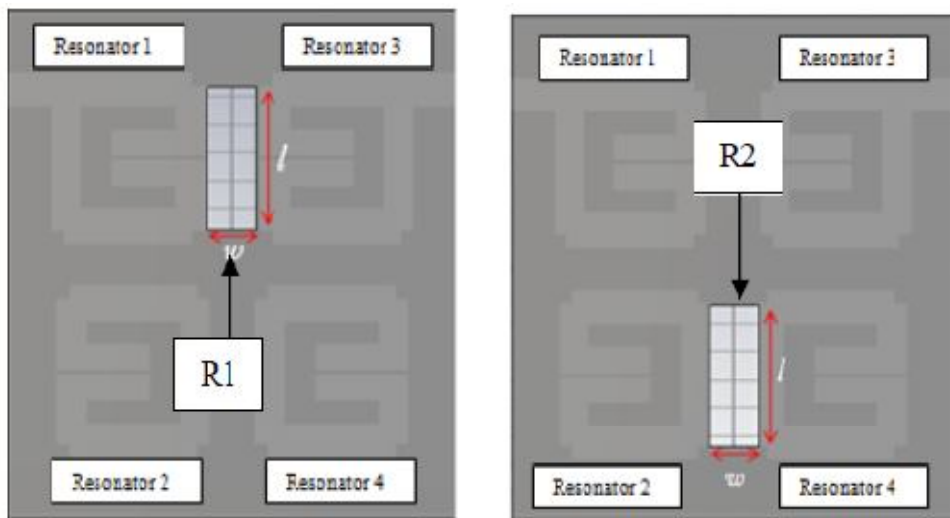


Figure 5: Proposed DGS Slot Layout

4. Simulation Results and Analyses

Simulations were performed using electromagnetic simulation software. Optimisations were carried out for fine-tuning in order to obtain an optimum configuration of the filter with integrated DGS elements. Two sets of QE filters are designed and simulated. The first set is without

DGS, while the second set is with variations of rectangular DGS slot elements. The filters are as follows:

- i. Quasi elliptic Bandpass Filter without DGS Element, QE
- ii. QE filter with first rectangular DGS of electric coupling, QE_R1
- iii. QE filter with second rectangular DGS of magnetic coupling, QE_R2
- iv. QE filter with two rectangular DGS Slots, QE_2R

4.1 QE Filter

The simulated results of the QE filter are shown in Figure 6. It can be observed that the filter showed elliptic response. It has central frequency of 2.4 GHz, and operating bandwidth of 79.15 MHz. Slight ripple of insertion loss is observed above the -3 dB level. Return loss is between -5 dB and -23.5dB within the passband region, indicating good impedance match at the input. The region occurs slightly lower than -3 dB level. Two close transmission zeros are observed at the lower and higher bands. Within the observed window, the stopband response is below -30 dB, indicating excellent stopband. The transition bands are almost symmetrical.

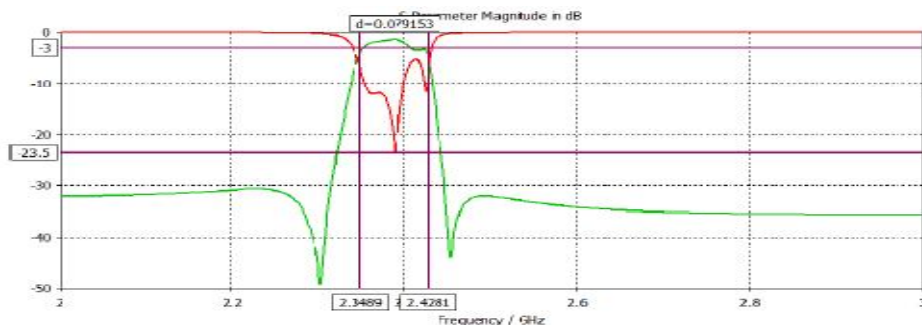
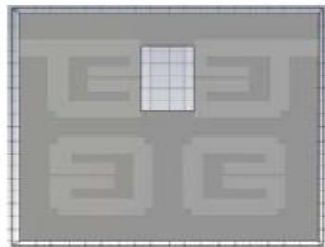


Figure 5: Simulated responses of QE filter

4.2 QE_R1 Filter

Figure 6 shows the geometrical layout and dimensions of the R1 DGS slot. Simulated responses are shown in Figure 7. It can be observed that good simulated return loss of better than -15 dB is obtained over the passband region, which is better than QE filter. Excellent return loss of -31.23 is obtained at 2.42 GHz. The filter operates from 2.37 GHz to 2.45 GHz, with a bandwidth of 85.71 MHz. Two close transmission zeros are observed at the lower and higher band. Within the observed window, the stopband response is below -20 dB, indicating good stopband. The transition bands are less symmetrical compared to QE filter.



<i>DGS</i>	<i>Length (mm)</i>	<i>Width (mm)</i>
<i>R1</i>	<i>5</i>	<i>2</i>

Figure 6: Geometrical Layout and Dimension of R1 DGS Slot

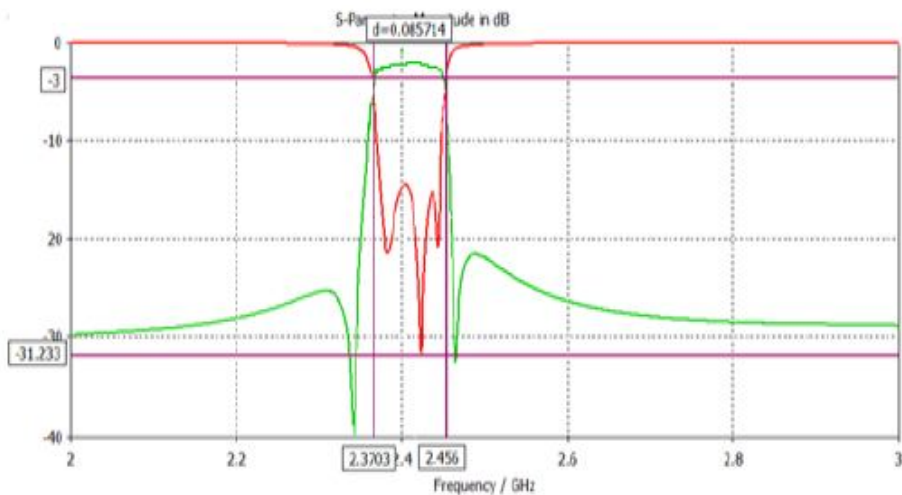
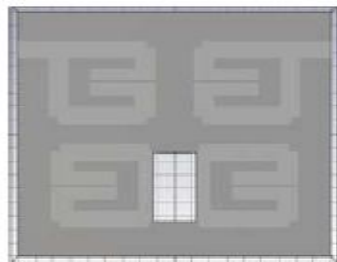


Figure 7: Simulated Responses of QE_R1 Filter

4.3 QE_R2 Filter

The geometrical layout and dimensions of the R2 DGS slot are shown in Figure 8. Figure 9 shows the simulated responses. It can be seen that sufficient passband return loss below -5 dB is observed. It records return loss of -14.87 dB at 2.4 GHz. The -3 dB bandwidth of the filter is 105.49 MHz, which is from 2.36 GHz to 2.47 GHz. The value is 25 MHz broader than QE_R1 filter. Stopband responses are below -30 dB at the lower and higher bands. Although better than that of QE_R1 filter, its return loss within the passband region is worse. This could be due to the effect of weak magnetic coupling compared to the electric coupling, contributed by the DGS slot.



<i>DGS</i>	<i>Length (mm)</i>	<i>Width (mm)</i>
<i>R2</i>	<i>5</i>	<i>2.4</i>

Figure 8: Geometrical Layout and Dimensions of R2 DGS Slot

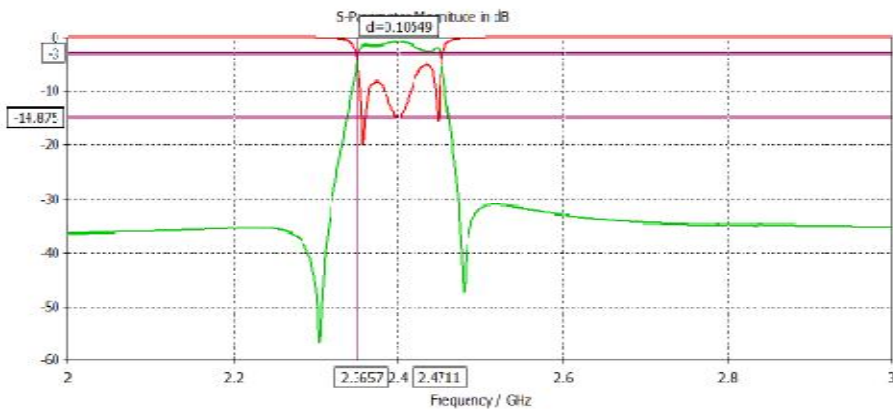
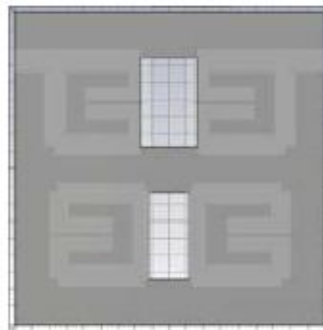


Figure 9: Simulated Responses of QE_R2 Filter

4.4 QE_2R Filter

In order to improve the stopband performance of the QE filter using one DGS slot, the filter is modified by having two DGS units. For QE_R1 filter, optimum R1 size is 10 mm² while for QE_R2 filter, optimum size is 12 mm². Figure 10 shows the geometrical layout and dimensions of the two proposed DGS slots.

Simulated responses are shown in Figure 11. It can be seen that the QE_2R filter is operating well at a centre frequency of 2.4 GHz, as desired with excellent good return loss of -28.07 dB and narrower operating bandwidth of 98.39 MHz or 4.5 %. The corresponding VSWR is 1.15. The value is approximately 6 MHz narrower than QE_R1 filter, but 18.39 MHz broader than QE_R2 filter. Stopband responses are similar ie -30 dB within the observed lower and higher bands. Its return loss within the passband region is better than -12 dB, which is sufficiently obtained. This could be due to the combination effect of the weaker magnetic coupling and stronger electric coupling, contributed by the DGS slots.



<i>DGS</i>	<i>Length (mm)</i>	<i>Width (mm)</i>
<i>R1</i>	5	2
<i>R2</i>	5	2.4

Figure 10: Geometrical Layout and Dimensions of 2R DGS slots

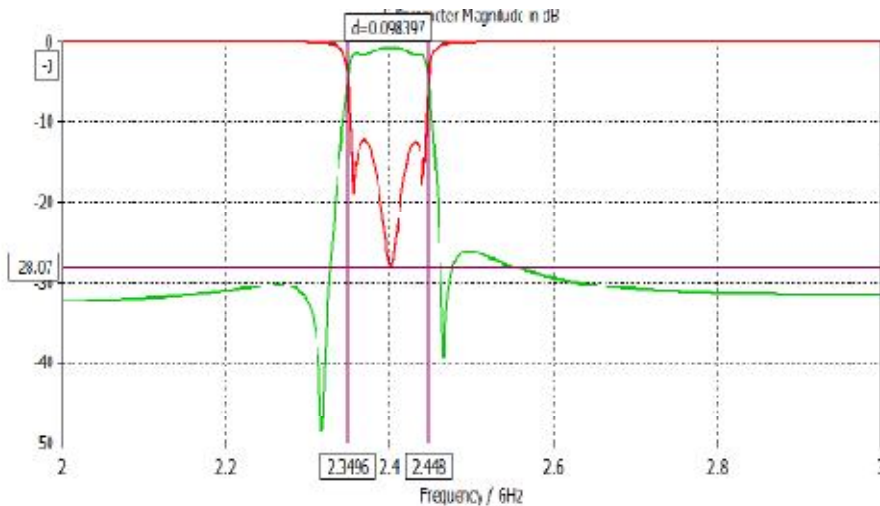


Figure 11: Simulated Responses of QE_2R Filter

5. Overall Discussion

The performance parameters of the four bandpass filter are tabulated and compared in Table 3. All bandpass filters operate well at similar operating frequencies, close to the desired 2.4 GHz, with excellent return losses of below -10 dB. The proposed QE_2R bandpass filter structure, employing the two DGS slots, has a wider fractional bandwidth of approximately 4.5 % at the centre frequency of 2.4 GHz compared to 3.3% for the conventional QE filter which is 19.39 MHz wider. This is because of the effect from the DGS structures that give tight couplings between adjacent resonators.

The increase of couplings in the presence of a ground slot has a simple physical explanation (*N. Militaru 2007*). For a conventional microstrip structure in the electric coupling configuration, many of the electric lines begin from a resonator end on the ground plane. With the presence of the slot, part of the lines is forced to end on the other resonator, tightening the electric coupling, or the electric part of a mixed coupling. QE_2R also shows slightly better performance in terms of return loss which is approximately -5 dB better than QE filter that is -23.5 dB, at 2.4 GHz. Both indicate very small reflections of signal at the input, indicating that the filters have well matched input and transmitting over 90% of signal power.

Table 3: Summary of Simulated Performance

Filter	Centre frequency, f_c (GHz)	Bandwidth, BW (MHz)	Return loss, S11 (dB)	VSWR
QE	2.4	79	-23.5	1.2
QE_R1	2.41	80	-30	1.13
QE_R2	2.4	105	-14.8	1.45
QE_2R	2.4	98.39	-28.07	1.15

6. Conclusion

This article presents a study on the effect of filter performance by implementing DGS element and investigating different geometrical structures. Two sets of QE filters have been successfully designed and investigated. The first set is without DGS slot, while the second set is with variations of rectangular DGS slot elements. Simulations results showed that the presence of DGS slot can provide better performance parameters. It can be concluded that the QE_2R filter exhibits excellent low return loss of -28.07 dB at 2.4 GHz, and the bandwidth of the filter reached to 98.39 MHz which is 19.39 MHz wider than the QE filter. Its stopband responses are similar ie -30 dB within the observed lower and higher bands.

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